BARR & STROUD LIMITED

ELECTRONIC FILTER SYSTEM EF3

Incorporating Filter Units

EF3-01 (high-pass) and EF3-02 (low-pass) EF3-03 (high-pass) and EF3-04 (low-pass)

TECHNICAL HANDBOOK



ELECTRONIC FILTER SYSTEM EF3



Incorporating Filter Units

EF3-01 (high-pass) and EF3-02 (low-pass) EF3-03 (high-pass) and EF3-04 (low-pass)

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SECTION 1 - SYSTEM DATA

1. Introduction

Electronic Filter System EF3 creates the opportunity to build up filter instrumentation by the most economical and convenient method - interchangeable filter units which plug into a basic power unit.

The low-profile cabinet, contains the plug-in power unit and up to two plug-in filter units. These can be switched to operate individually, or in cascade, or in other combinations to provide filtering modes such as band-pass, band-stop, band-separate and band-combine.

Technical details of plug-in filter units currently available are given in subsequent sections of this handbook.

2. Description

The cabinet has two bays which accommodate the selected plug-in filter units. The units locate easily on fixed rails with connections completed by plug and socket breaks. Each unit is secured in the cabinet by two fasteners.

The cabinet is designed for bench mounting, and has an integral folding stand to facilitate operation of controls. The cabinet can be adapted for 19-inch rack mounting, if required, using conversion kit No. EF3954 (see Figure 1).

Weight of cabinet with power unit and two filter units is 4.5kg.

A. Power Unit

The power unit can be mains or battery operated, and provides stabilised 15V positive and negative supplies to the filter units. Mains and battery connectors are on the rear. The mains ON/OFF switch on the front of the unit does not control the battery supplies.

The supply requirements are:

Mains:

205 to 245V (110V tap), 25W, 50/60 Hz, maximum mains

variation \pm 7% on voltage.

Battery:

+24V and -24V at 100mA per filter unit, maximum battery

variation + 15% on voltage.

The 15V positive and negative lines are separately fused (250mA). The fuses are on the printed circuit board within the power unit. A spare fuse is also present on the board. The mains transformer is protected by a 250mA fuse accessible from the rear of the instrument.

The battery sockets are permanently connected across the bridge rectifier and are therefore 'live' while the unit is being operated by mains supply. Two diodes are connected in series with the battery sockets to protect against inadvertent external 'shorting' of the sockets when using a mains supply. The diodes also protect against accidental reversal of battery polarities.

The power unit circuit diagram is shown on Figure 5 (EF3-16 power units) or Figure 5A (EF3-17 power units).

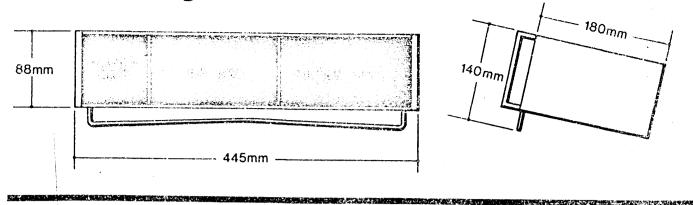
B. Environmental conditions

Operating temperature range
Storage temperature range

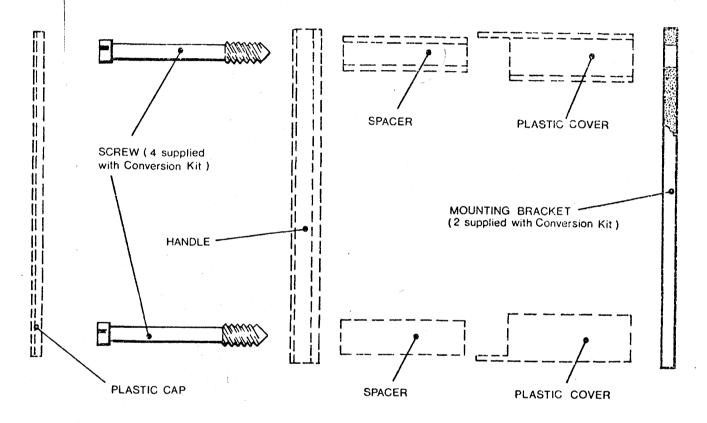
 0° C to + 45 $^{\circ}$ C

 -20° C to $+80^{\circ}$ C

Dimensions Diagram



Conversion Kit No. EF3954 (for 19-inch rack mounting)



Note:

Items with solid lines are the components of Conversion Kit. No. EF3954. Items with broken lines are components of the handle.

Instructions:

- (1) Remove plastic cap from handle.
- (2) Unscrew and remove the two handle screws.
- (3) Position mounting bracket between cabinet and handle spacer, aligning bracket holes with those in cabinet.
- (4) Secure handle using two conversion kit screws.
- (5) Fit plastic cap.

Dimensions Diagram and Conversion Kit Figure 1



SECTION 2 - PLUG-IN FILTER UNITS

High-Pass Filter Unit EF3-01 1.

EF3-01 can be used in the basic cabinet with the power unit, either as a single filter unit or cascaded with a second EF3-01 unit for increased attenuation rate. More generally it is used with a low-pass filter unit from System EF3 to provide a full range of modes: band-pass, band-stop, band-separate, and band-combine.

The EF3-01 has a response, expressed in terms of maximum 3dB bandwidth from 0.01Hz to 500kHz. Cut-off frequency is variable from 0.01Hz to 10kHz and is selected by digital controls consisting of 2 decade switches and a 5 range multiplier. selection has the advantage of accurate repeatability of setting.

The filter response is 8-pole Butterworth for flattest passband response with a nominal final attenuation rate of 48dB/octave in the cut-off region. A damped characteristic can be selected for minimum distortion of complex waveforms.

Input and output connections are via BNC sockets on the front panel with parallel sockets at the rear. The unit is protected against accidental damage by output short circuit.

Specification

Maximum 3dB bandwidth	0.01Hz to 500kHz	
Cut-off frequency F _c variable from 0.01Hz to 10kH		
Calibration accuracy	$\pm 3\%$ on frequency setting	
Attenuation rate	48dB/octave }	
Passband insertion loss	$0 \pm 0.5 dB$	
Insertion loss at F _c (normal mode)	3 ± 0.5 dB $\begin{cases} see Figure 2 \end{cases}$	
Insertion loss at F _c (damped mode)	$14 \pm 2dB$	

Final attenuation	>85 dB
Passband limit	6dB/octave falling off from 500kHz to 1 MHz thereafter 12dB/octave
Phase response	see Figure 2
Maximum input signal	7V peak (5V rms)
Permissible d.c. component	150Vmaximum at input
Maximum output current	normally 20mA (guaranteed 10mA)
Harmonic distortion	<0.2% below 500kHz
Input impedance	$4 M\Omega$ in parallel with $60 pF$
Output impedance	50Ω
Offset d.c. drift/time	typically <u>+</u> 1m V/day after 2 hour warm - up period
Offset d.c. drift/temp.	typically <100μV/C after

Change in offset d.c. volts

25mV maximum over the whole

2 hour warm-up period

F_c range



Warm-up period

5 minutes for maximum a.c. signal handling

Noise level

 $300\mu V \, \mathrm{rms}$ over $500 \, \mathrm{kHz}$ bandwidth, with input short circuited (battery or mains operated)

All values stated are nominal unless tolerances are specified.

B. Circuit description

EF3-01 filter unit has been synthesised using R-C ladder networks in conjunction with operational amplifiers. The design method used permits the realisation of an 8-pole characteristic by cascading four 2-pole sections.

Figure 6 shows the circuit detail of the unity gain amplifiers on the plug-in amplifier board and the pin connections from the edge connector (SK2) to the R-C switch sections. Details of the R-C switch sections are shown on Figure 8.

R46 is adjusted to set the gain of X8.

R10 equalises the insertion loss on the x 100 range of the high-pass unit. On all other high-pass ranges the insertion loss is constant and R10 is not in circuit.

R20 and R28 are adjusted to provide the correct feedback to ensure that the frequency response is within the prescribed limits.

X8, the output amplifier, has greater than unity gain (18dB approx.) to compensate for losses in the remainder of the network.

2. Low-pass Filter Unit EF3-02

EF3-02 can be used in the basic cabinet with the power unit, either as a single filter unit or cascaded with a second EF3-02 unit for increased attenuation rate. More generally it is used with a high-pass filter unit from System EF3 to provide a full range of modes: band-pass, band-stop, band-separate, and band-combine.

The EF3-02 has a maximally flat response from d.c. to 10 kHz. Cut-off frequency is variable from 0.01Hz to 10 kHz and is selected by digital controls consisting of 2 decade switches and a 5 range multiplier. Digital selection has the advantage of accurate repeatability of setting.

The filter response is 8-pole Butterworth for flattest passband response with nominal final attenuation rate of 48dB/octave in the cut-off region. A damped characteristic can be selected to provide linear phase response for minimum distortion of complex waveforms.

Input and output connections are via BNC sockets on the front panel with parallel sockets at the rear. The unit is protected against accidental damage by output short-circuit.

A. Specification

Maximum bandwidth

Cut-off frequency F

Calibration accuracy

Attenuation rate

Passband insertion loss

d.c. to 10kHz (3dB down)

variable from 0.01Hz to 10 kHz

+3% on frequency setting

48dB/Octave

0 + 0.5 dB

see Figure 3

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Insertion loss at F _c (normal mode)	$3 \pm 0.5 dB$
Insertion loss at F _c (damped mode)	$ \begin{cases} \text{see Figure 3} \\ 14 \pm 2 \text{dB} \end{cases} $
Final attenuation	>75 dB to 2 MHz (min)
Phase & delay response	see Figure 3
Square-wave response	see Figure 3
Maximum input signal	7V peak (5V rms) or 7V d.c. (combined a.c.and d.c.components of input must not exceed 7V peak)
Maximum output current	normally 20mA (guaranteed 10mA)
Harmonic distortion	<0.2% below 10kHz
Input impedance	4 Μ Ω in parallel with 60 pF
Output impedance	50Ω
Offset d.c. drift/time	typically + lmV/day after 2 hour warm-up period
Offset d.c. drift/temp.	typically < $100 \mu \text{ V} / ^{\text{O}}\text{C}$ after 2 hour warm-up period
Change in offset d.c. volts	25 mV maximum over the whole F range
Noise level	200μV rms over 10kHz bandwidth, with input short-circuited (battery or mains operated)

All values stated are nominal unless tolerances are specified.

B. Circuit description

EF3-02 filter unit has been synthesised using R-C ladder networks in conjunction with operational amplifiers. The design method used permits the realisation of an 8-pole characteristic by cascading four 2-pole sections.

Figure 6 shows the circuit detail of the unity gain amplifiers on the plug-in amplifier board and the pin connections from the edge connector (SK 2) to the R-C switch sections. Details of the R-C switch sections are shown on Figure 8.

R46 is adjusted to set the gain of X8.

 $R\,20$ and $R\,28$ are adjusted to provide the correct feedback to ensure that the frequency response is within the prescribed limits.

X8, the output amplifier, has greater than unity gain (18dB approx.) to compensate for losses in the remainder of the network.



3. High-pass Filter Unit EF3-03

EF3-03 can be used in the basic cabinet with the power unit, either as a single filter unit or cascaded with a second EF3-03 unit for increased attenuation rate. More generally it is used with a low-pass filter unit from System EF3 to provide a full range of modes band-pass, band-stop, band-separate, and band-combine.

The EF3-03 has a response, expressed in terms of maximum 3 dB bandwidth from 0.1Hz to 700kHz. Cut-off frequency is variable from 0.1 Hz to 100kHz and is selected by digital controls consisting of 2 decade switches and a 5 range multiplier. Digital selection has the advantage of accurate repeatability of setting.

The filter response is 8-pole Butterworth for flattest passband response with a nominal final attenuation rate of 48dB/octave in the cut-off region. A damped characteristic can be selected for minimum distortion of complex waveforms.

Input and output connections are via BNC sockets on the front panel with parallel sockets at the rear. The unit is protected against accidental damage by output short-circuit.

A. Specification

Maximum 3dB bandwidth

Cut-off frequency F

Calibration accuracy

Attenuation rate

Passband insertion loss >

Insertion loss at F_c (normal mode)

Insertion loss at F_c (damped mode)

Final attenuation

Passband limit

Phase response

Maximum input signal

Permissible d.c. component

Maximum output current

Harmonic distortion

Input impedance

Output impedance

0.1 Hz to 700 kHz

variable from 0.1Hz to 100kHz

+ 3% on frequency setting

48dB/octave

 0 ± 0.5 dB

 3 ± 0.5 dB

see Figure 2

16 + 2 dB

>85 dB

6dB/octave falling off from 700kHz to 1.5MHz thereafter 12dB/octave

see Figure 2

7V peak (5V rms) up to 300kHz reducing to 2.5V peak at 700kHz for undistorted operation.

150V maximum at input

normally 20mA (guaranteed 10mA)

<0.2% below 700kHz

 $4M\Omega$ in parallel with 60 pF

 50Ω



Offset d.c. drift/time

typically + 1m V/day after 2 hour warm-up period

Offset d.c. drift/temp.

typically <100 μ V/ 0 C after 2 hour warm-up period

Change in offset d.c. volts

3mV over the whole F range

Noise level

350μ V rms over 700kHz bandwidth, with input short circuited (battery or mains operated).

All values stated are nominal unless tolerances are specified.

B. Circuit description

EF3-03 filter unit has been synthesised using R-C ladder networks in conjunction with operational amplifiers. The design method used permits the realisation of an 8-pole characteristic by cascading four 2-pole sections.

Figure 9 shows the circuit detail of the unity gain amplifiers on the plug-in amplifier board and the pin connections from the edge connector (SK2) to the R-C switch sections. Details of the R-C switch sections are shown on Figure 8.

R10 is adjusted to equalise the insertion loss on the x 1K range of the high-pass unit. On all other high-pass ranges the insertion loss is constant and R10 is not in circuit.

R20 and R28 are adjusted to provide the correct feedback to ensure that the frequency response is within the prescribed limits.

X8, the output amplifier, has greater than unity gain (18dB approx.) to compensate for losses in the remainder of the network.

R46 is adjusted to set the gain of X8.

4. Low-pass Filter Unit EF3-04

EF3-04 can be used in the basic cabinet with the power unit, either as a single filter unit or cascaded with a second EF3-04 unit for increased attenuation rate. More generally it is used with a high-pass filter unit from System EF3 to provide a full range of modes: band-pass, band-stop, band-separate, and band-combine.

The EF3-04 has a maximally flat response from d.c. to 100kHz. Cut-off frequency is variable from 0.1Hz to 100kHz and is selected by digital controls consisting of 2 decade switches and a 5 range multiplier. Digital selection has the advantage of accurate repeatability of setting.

The filter response is 8-pole Butterworth for flattest passband response with nominal final attenuation rate of 48dB/octave in the cut-off region. A damped characteristic can be selected to provide linear phase response for minimum distortion of complex waveforms.

Input and output connections are via BNC sockets on the front panel with parallel sockets at the rear. The unit is protected against accidental damage by output short-circuit.



A. Specification

Maximum bandwidth

Cut-off frequency F

Calibration accuracy

Attenuation rate

Passband insertion loss

Insertion loss at F (normal mode)

Insertion loss at F_c (damped mode)

Final attenuation

Phase & delay response

Square-wave response

Maximum input signal

Maximum output current

Harmonic distortion

Input impedance

Output impedance

Offset d.c. drift/time

Offset d.c. drift/temp.

Change in offset d.c. volts

Noise level

d.c. to 100kHz (3dB down)

variable from 0.1Hz to 100kHz

+ 3% on frequency setting

48dB/octave

0 + 0.5 dB

3 + 0.5 dB

see Figure 3

 $16 \pm 2 \, dB$

 $>75\,\mathrm{dB}$ up to $5~\mathrm{MHz}$

>60 dB up to 10 MHz

see Figure 3

see Figure 3

7V peak (5V rms) or 7V d.c.

(combined a.c. and d.c. components of input must not exceed 7V peak) for

undistorted operation

normally 20mA

(guaranteed 10mA)

<0.2% below 100kHz

 $4M\Omega$ in parallel with 60pF

 50Ω

typically $\pm 1 \text{mV/day after}$

2 hour warm-up period

typically $<100\mu\text{V/}^{0}\text{C}$ after

2 hour warm-up period

3mV over the

whole F range

250µV rms over 100kHz bandwidth,

with input short-circuited (battery

or mains operated)

All values stated are nominal unless tolerances are specified.



B. Circuit description

EF3-04 filter unit has been synthesised using R-C ladder networks in conjunction with operational amplifiers. The design method used permits the realisation of an 8-pole characteristic by cascading four 2-pole sections.

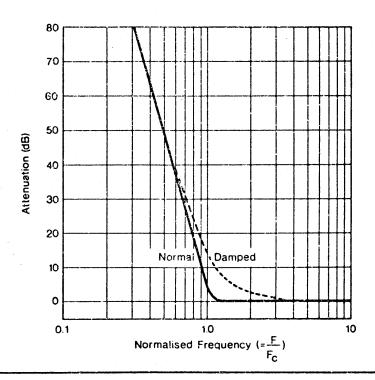
Figure 9 shows the circuit detail of the unity gain amplifiers on the plug-in amplifier board and the pin connections from the edge connector (SK2) to the R-C switch sections. Details of the R-C switch sections are shown on Figure 8.

R 20 and R 28 are adjusted to provide the correct feedback to ensure that the frequency response is within the prescribed limits.

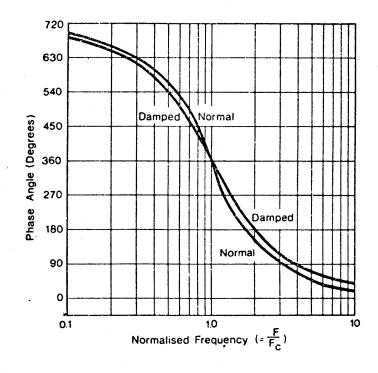
X8, the output amplifier, has greater than unity gain (18dB approx.) to compensate for losses in the remainder of the network.

R46 is adjusted to set the gain of X8.

Attenuation



Phase

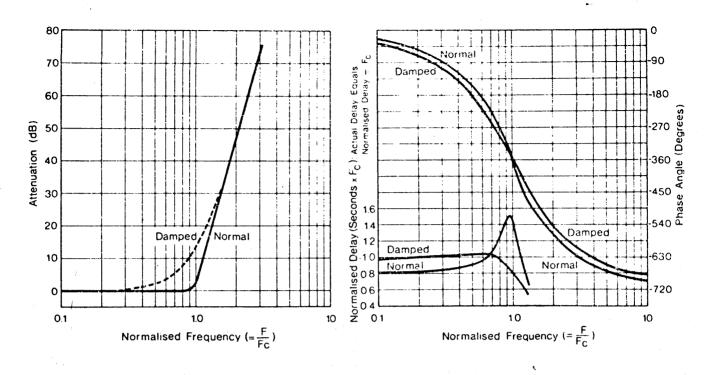


Response: High-Pass Filter Units EF3-01 & EF3-03

Figure 2

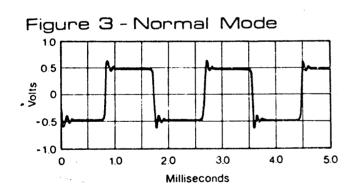
Attenuation

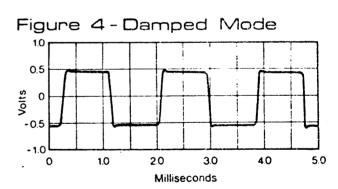
Phase & Delay



Typical Square wave Response

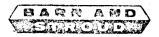
Filter unit switch setting, $F_c = 10kHz$, 500Hz square-wave input.





Response: Low-Pass Filter Units EF3.02 & EF3.04

Figure 3



SECTION 3 - LINKED OPERATION

1. Filter Units EF3-01(high-pass) & EF3-02(low-pass); EF3-03(high-pass) & EF3-04(low-pass)

These filter units provide the following range of operational modes:

- (a) Isolate two filter units available for independent use.
- (b) Band-Pass
- (c) Band-Stop
- (d) Band-Separate
- (e) Band-Combine

one high-pass unit & one

low-pass unit interconnected

Appropriate interconnections are made by the mode selection switch on the power unit. Signal connections are shown in the diagrams.

Cascade - two filter units of the same type, connected in series.

NOTE:

(f)

All values stated are nominal unless tolerances are specified.

A. Isolate

This mode permits independent use of two filter units.

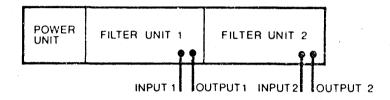
B. Band-Pass

Attenuation slopes outside the passband are 48dB/octave.

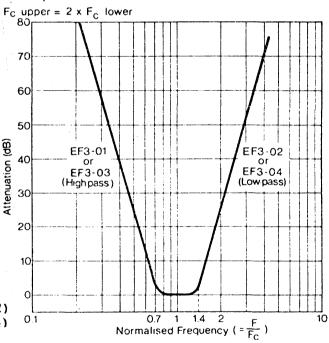
Passband insertion loss is dependent on the ratio between the upper and lower cut-off frequencies as follows:

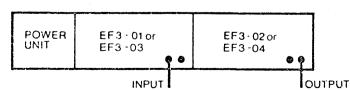
- (a) When $\frac{\text{Fc upper}}{\text{Fc lower}} \ge 1.6$, insertion loss = 0 + 1 dB.
- (b) When Fc upper approaches 1.0, insertion loss increases to approximately 6dB(normal mode) 28dB (damped mode EF3-01 & EF3-02) 32dB (damped mode EF3-03 & EF3-04)

Input impedance is $4M\Omega$ in parallel with 60 pF.



Response (Normal Mode)







C. Band-Stop

Attenuation slopes are 48dB/octave.

Midband frequency attenuation is dependent on the ratio between the upper and lower cut-off frequencies as follows:

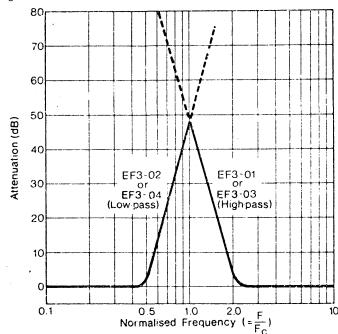
- (a) When $\frac{\text{Fc upper}}{\text{Fc lower}}$ attenuation $\geq 42 dB$.
- Fc upper approaches 1.0, (b) When attenuation is decreased to approximately OdB (normal mode) 10dB (damped mode).

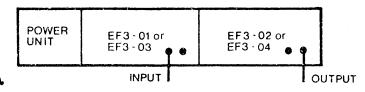
Input impedance is $2M\Omega$ in parallel with 120 pF.

The output amplifier of the right-hand filter unit compensates for the summing losses in R35 and R42 (see circuit diagrams Figs. 6 & 9).

Response (Normal Mode)

 F_C upper = $4 \times F_C$ lower

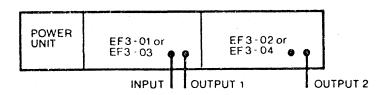




D. Band-Separate

Attenuation slopes are 48dB/octave.

Input impedance is $2M\Omega$ in parallel with 120 pF.

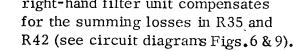


E. Band-Combine

Attenuation slopes are 48dB/octave.

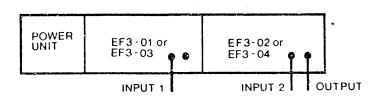
Input impedance for each unit is nominally 4M\(\Omega\) in parallel with 60 pF.

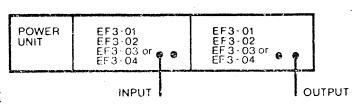
The output amplifier of the right-hand filter unit compensates



F. Cascade

Insertion loss at the selected cut-off frequency is approximately 6dB (normal mode) 28dB (damped mode EF3-01 & EF3-02) 32dB (damped mode EF3-03 & EF3-04)





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SECTION 4 - OPERATION

1. Operating Controls

A. Frequency selection switches

Two frequency selection switches and a multiplier switch on each filter unit determine the cut-off frequency (in the normal mode the 3 dB attenuation frequency is commonly referred to as the cut-off frequency). The left-hand frequency selection switch is calibrated in 'tens' and the other is calibrated in 'units'. Cut-off frequency is derived from the sum of the readings on the 'tens' and 'units' switch multiplied by the multiplier switch setting.

B. Multiplier switch

This switch has five positional settings providing multiplying factors of 0.01, 0.1, 1.0, 10 and 100, (EF3-01 & EF3-02), and 0.1, 1.0, 10, 100, 1K, (EF3-03 & EF3-04).

C. Response switch

The response switch controls the form of response provided by the unit. It has two settings as follows:

- (i) Normal which provides a response similar to that of an 8-pole Butterworth function and therefore, has the flattest monotonic passband response possible; it has 3 dB attenuation at the cut-off frequency, thereafter increasing at the rate of 48dB/octave.
- Damped which provides an improved filter phase response, effectively reducing ringing and overshoot on pulse and step-type waveforms. This is particularly useful in the low-pass unit. Attenuation at the selected cut-off frequency increases to approximately 14dB(EF3-01 & EF3-02) 16dB (EF3-03 & EF3-04); the 3dB frequency is at approximately 0.5 x the selected frequency in the low-pass unit and 2 x the selected frequency in the high-pass unit.

D. d.c. zero control

output of the left-hand unit.

This enables the output of the unit to be set to zero volts d.c.

E. Mode switch

This is situated on the power unit and enables the two filter units to be used as individual or combined units depending on the required mode of operation. The mode switch has six positions which are:

- (i) Isolate The two filter units are isolated from one another both at the input and output, but share the power supply.
- Band-Pass The output of the left-hand unit is connected to the input of the right-hand unit and provides band-pass facilities when one high-pass and one low-pass unit is used (it is usual to insert the high-pass unit in the left-hand position to eliminate any d.c. component present in the input). The high-pass unit determines the lower band-edge frequency and the low-pass unit the upper band-edge frequency. Input connection is made to the left-hand unit and bandpass output taken from the right-hand unit.

 Signals present in the stop band of the right-hand unit are available at the

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- NOTE In band-pass operation if both units are set to the same cut-off frequency a $6 \pm 1 dB$ loss will occur at the mid-band with the response switch in the 'normal position'.
- (iii) Band-Stop One high-pass and one low-pass unit are required for this mode. The inputs and outputs of the two units are connected in parallel. Input may be connected to either input socket but output containing both passbands is available from the right-hand unit only.
- (iv) Separate The inputs of the two units are connected in parallel but the outputs are independent of one another. Input signal may be connected to either input socket.
- (v) Combine The inputs of the two units are independent of each other and will therefore accept two separate signals. The combined output is available at the right-hand unit only.
- (vi) Cascade The output of the left-hand unit is connected to the input of the right-hand unit to enable two low-pass or two high-pass units to be used to obtain an increased cut-off rate. Alternatively, two different high-pass or low-pass bandwidths can be obtained by setting the left-hand unit to the wider band.

2. Setting Up

A. Power supply connection

System EF3 can be powered by mains or battery supply. Before connecting to a mains supply, it is essential that the correct voltage is selected at the mains transformer (the instrument is despatched with the transformer adjusted for 240V a.c., 50/60 Hz mains supply).

Mains supply connection is made to a socket on the rear of the power unit by a 3-core cable provided with the instrument. Wire identification is as follows: brown - line; blue - neutral; green/yellow - ground. Battery supply connection is made to three 4mm sockets in the rear of the power unit. These sockets are colour coded as follows: red - +24V; yellow- -24V; black - common or ground.

B. Set d.c.zero at output

Proceed as follows:

(a) High-pass unit

- (i) With input open-circuited, connect a suitable d.c. voltmeter to the output.
- (ii) Adjust the preset control R38 (d.c. zero) for zero on the meter.

NOTE Because the input is capacitively coupled, short-circuiting of the input is not required.

(b) Low-pass unit

- (i) With input short-circuited, connect d.c. voltmeter to output.
- (ii) Adjust the preset control R38 (d.c. zero) for zero on the meter.



(iii) Remove short-circuit from input.

NOTE Small changes in the d.c. offset at the filter output may occur when the digital switch settings are changed. Maximum variation over the whole range of settings is 25 mV (EF3-01 & EF3-02) and 3mV (EF3-03 & EF3-04).

C. Signal connection

Connect signal to the appropriate sockets as determined by the required filtering mode. The diagrams in Section 3 indicate the various connection arrangements.

3. Operating Procedure

The operating procedure consists of correctly setting the two frequency selection switches and the multiplier switch.

e.g.	'Tens' Switch Setting	$2 = 20 (i.e. 2 \times 10)$
	'Units' Switch Setting	5
	Sum of Switch Settings	25
	Multiplier Switch Setting	x 10
	Cut-off Frequency	$25 \times 10 = 250$ Hz

NOTE When filters are used independently of one another in the same frame the level of isolation between the units is determined in some measure by the source impedances. If the source impedance connected to the input is less than $10 \mathrm{K}\Omega$ then the isolation exceeds $60 \mathrm{dB}$. For $100 \mathrm{K}\Omega$ source this reduces to $50 \mathrm{dB}$ approximately and on open circuit to approximately $20 \mathrm{dB}$ down at the output of the unenergised unit. The level of isolation exceeds the stop band attenuation for source impedances less than $1.0 \mathrm{K}\Omega$.



SECTION 5 - FAULT FINDING

The following a.c. and d.c. voltages are given as an aid to fault location. These should be used for guidance only.

All voltages are measured with respect to Pin 1.

Measure the a.c. voltages as follows:

- (1) Apply a 2V a.c. signal to the input.
- (2) Set response switch to 'normal'.
- (3) Set the filter cut-off so that the test signal frequency is well within the passband, e.g. approximately 1/5 the cut-off frequency for low-pass or 5 times the cut-off frequency for high-pass.
- (4) Set mode switch to 'isolate'.
- Input and subsequent measurements may be either peak or rms as appropriate.
 - The x 100 range in EF3-01 and the x 1K range in EF3-03 should not be used for (ii) checking as the a.c. voltages on these ranges differ from those listed in the tables.

Table 1 - Edge Connector SK2; EF3-01/EF3-02/EF3-03/EF3-04

PIN No.	EF3-01/-02/-03/-04	EF3-01/-02/-03/-04
	a.c. VOLTS	d.c. VOLTS
1	ov	0V
. 2	ov	+15V <u>+</u> 0.5V (supply)
3	2. 0V	0V + 25mV
4	$1.0V \pm 50mV$	" "
5	ov –	-15V <u>+</u> 0.5V (supply)
6	ov	ōv
7	$1.0V \pm 50 \text{mV}$	0V + 25mV
8	" - " (-6dB)	11 11
9	" "	" ".
. 10	" "	, n
11	" " (-6dB)	" "
12	all in the second of the secon	n n
13	" " (-6dB)	" "
14	$0.34V \pm 50 \text{mV} (-16.48dB)$	" "
15	0V	$-2.5V \pm 0.5V$
16	$1.0V \pm 50 \text{mV}$	$0V \pm 25mV$
17	" " (-6dB)	" "
18		, , , , , , , , , , , , , , , , , , ,
19	" " (~6dB)	n n
20	0.94V + 50mV (-7.43dB	11 11
.21	" " < 296)	11 11
22	$1.0V \pm 50 \text{mV}$	11 11
23	" " (-6dB).	" "
24	11 11	" "
25 .	11 11	11 11
26	$0.19V \pm 50mV$	11 11
27	$1.0V \pm 50mV$	" "
28	0V	$+ 3.5V \pm 0.5V$
29	$2.0V \pm 50mV (0/P)$	0V
30	0V	0V <u>+</u> 1V
31	0V	0V
32	0V	0V



SECTION 6 - APPLICATION

In many systems it is necessary to introduce filtering to attenuate unwanted signals. The choice of a suitable filter network can be exceedingly difficult especially when it comes to specifying the required filter parameters. This task can be made easier by making use of a variable filter during the various stages of design. When the precise requirement has thus been established, fixed frequency passive or active networks can be designed and constructed in a suitable form for inclusion in the final equipment assembly. Barr & Stroud Limited design and produce such custom-built filter networks.

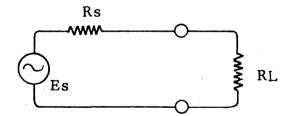
Consider the network shown on diagram A below, which represents a voltage source Es of resistance Rs and a resistive load termination RL. If **Es** is complex and some of its components are to be suppressed then a suitable filter must be introduced into the network as on diagram B. Such a filter would be designed to operate between source resistance Rs and load termination RL.

If a Variable Filter is inserted in such a network then the following points should be $noted_{\bullet}$

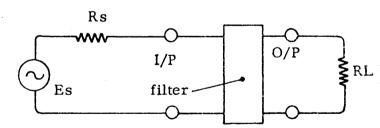
Because the input impedance of the variable filter is high a dummy load RL should be introduced at its input. This serves two purposes; firstly, the signal source will then operate under its normal loaded condition and, secondly, the insertion loss response of the variable filter will then be the same as that which would be obtained if it were replaced by an equivalent fixed frequency passive network with terminations Rs and RL. In the case where the load termination RL approaches the output impedance of the variable filter then a fixed increase in the insertion loss will result.

When the variable filter is being used it may not however be necessary to introduce it between the source and load as shown on diagram C. To observe the effect of filtering on the signal waveform the variable filter may be connected across the load RL as shown on diagram D provided RL is small compared with the filter input impedance. The waveform of the filter output is then the same as that which would be obtained were the filter inserted between source and load as on diagram C.

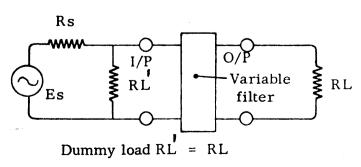
A. System to be filtered



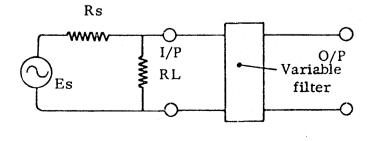
B. System with desired filter

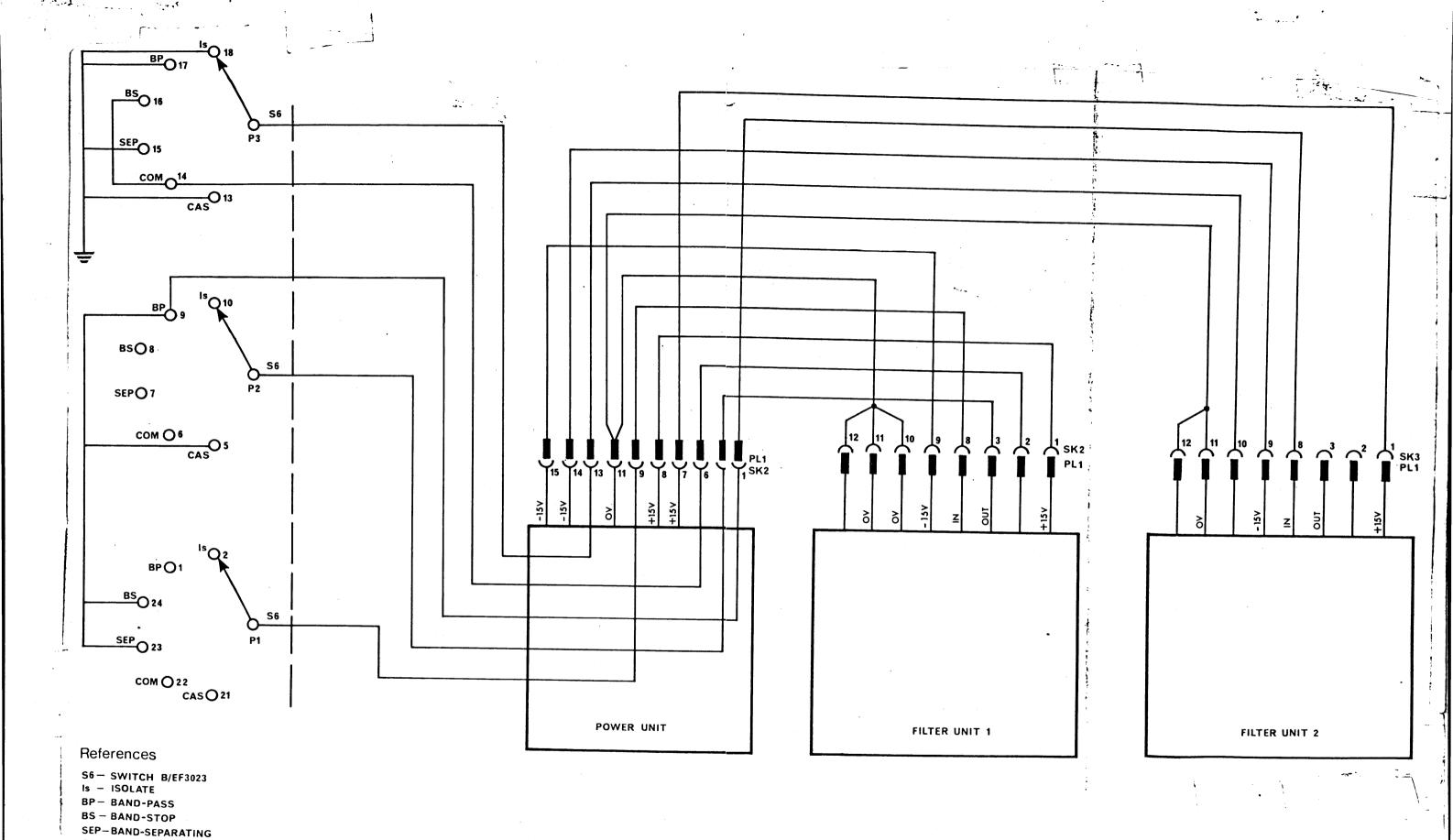


C. System with variable filter connected between source Rs & RL



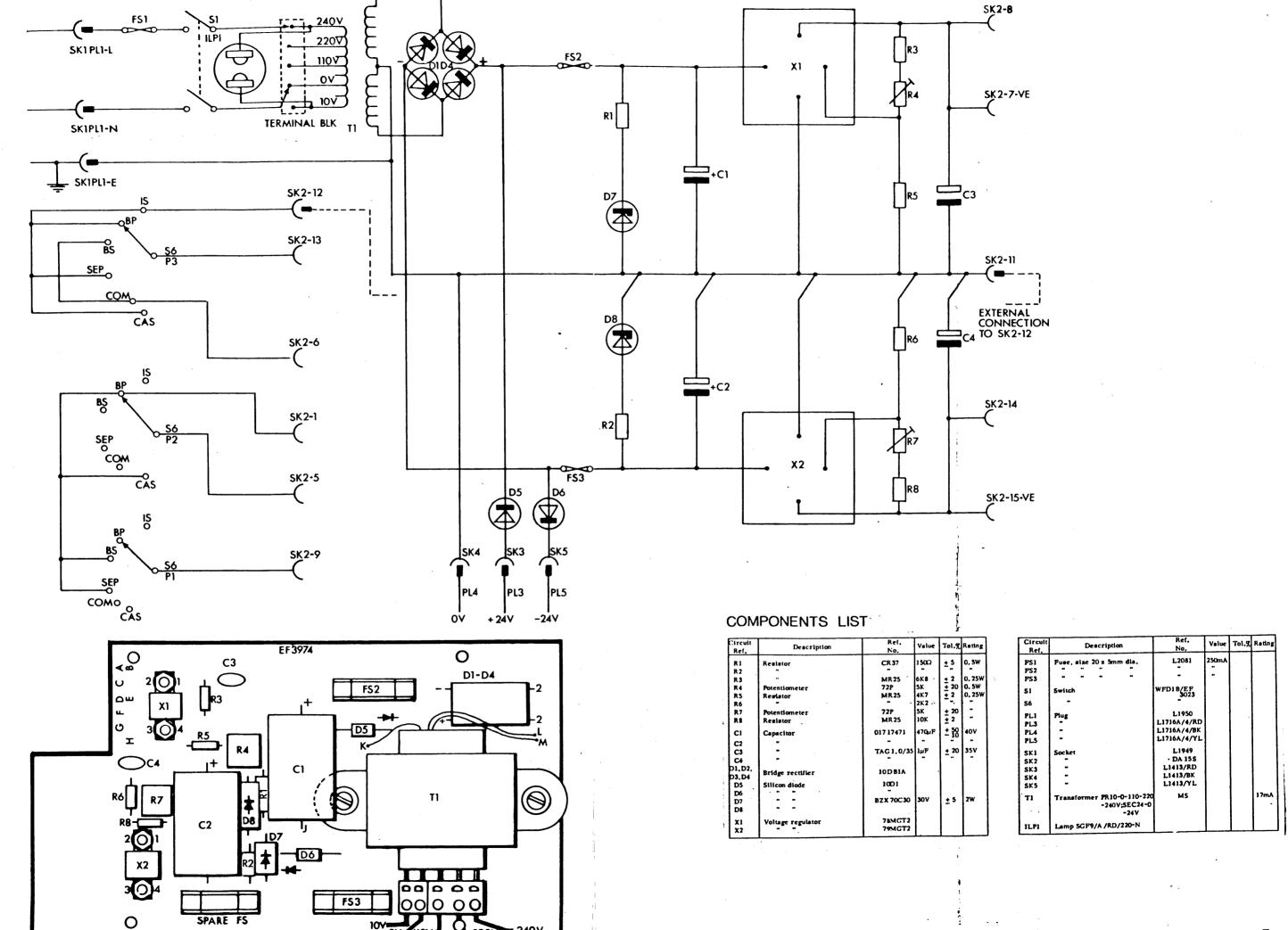
D. System with variable filter input connected across load RL





COM-BAND-COMBINING

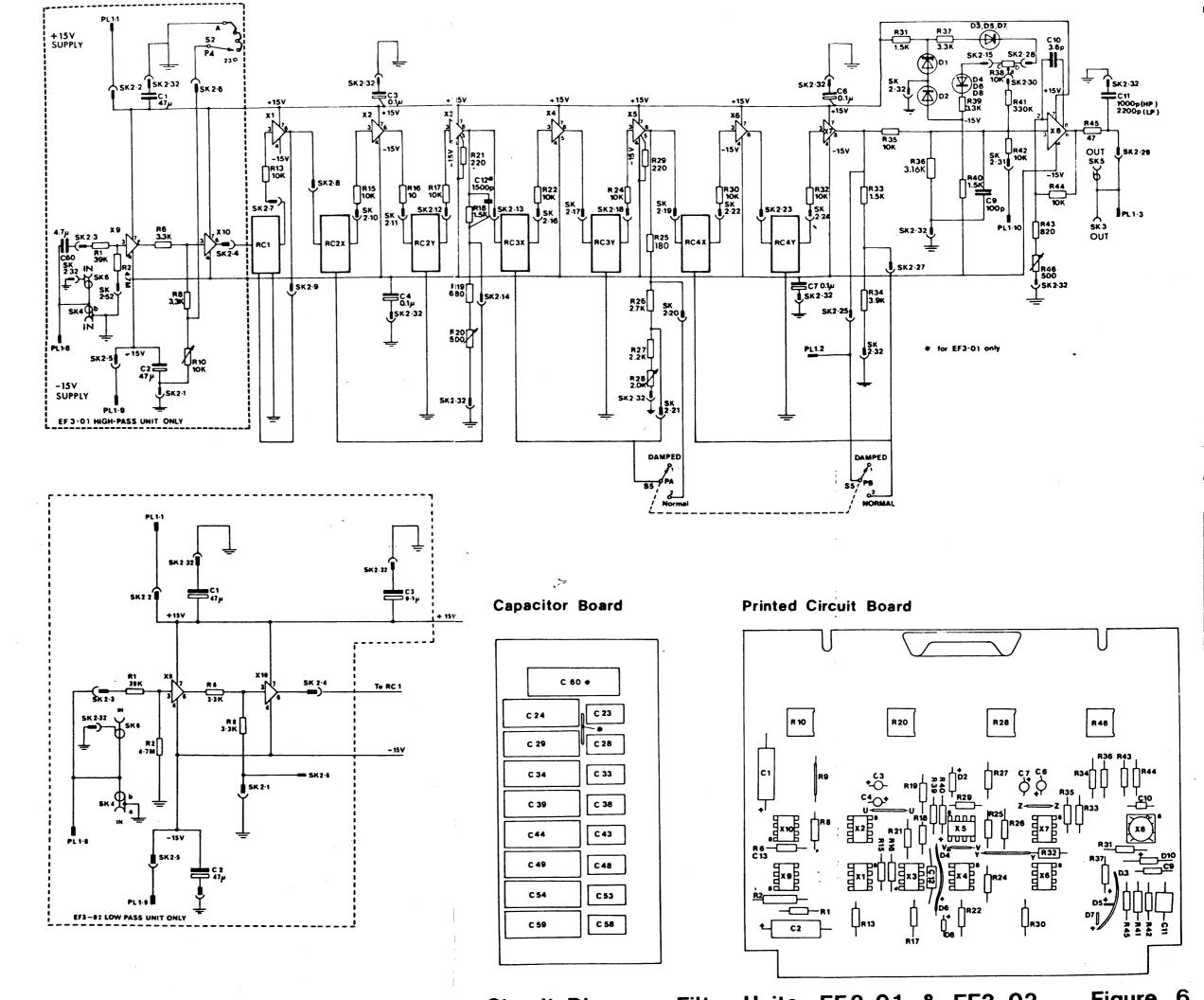
CAS-CASCADE



Circuit Diagram: Power Unit EF3-17

AUG 80

Figure 5



AUG 80

Circuit Diagram: Filter Units EF3-01 & EF3-02

Figure 6

Filter Units EF 3-01 (High-Pass) & EF 3-02 (Low-Pass)

Circuit Ref.	D	Ref.	Value	Tol."	Rating	Circuit	Description	Ref.	Value In	
Kei.	Description	No.	Ω			Ref.	Description .	No.	G,	ol. Rating
						R99 R100	Resistor, Code Z	ногсона		1 0.25 W
						R101			4871	1 0,25 W
R1	Resistor (Mullard)	CR 25	39K	± 5	0.33W	R102 R103			42.2 k	. 1 0.25 W
R2 R6	Resistor " Resistor (Mullard)	MR 25	4.7M	± 10	0 400	R104				1 0.25 W 1 0.25 W
R8	" " "	min 23	3.3K	± 2	0.4W	R105 R106			169 k	1 0.25 W
R13	* *	CR 25	10K	<u>+</u> 5	0.33W	1 R107			1 84 5 k 1 1	.1 0.25 W
R15 R16				"		R108 R109			68.1 k 56.2 k 48.7 k 42.2 k	1 0.25 W
R17			10Ω 10K			R110			56.2 k	.1 0.25 W .1 0.25 W
R18		-	1.5K	-		R111 R112			42.2 k	1 0.25 W
R19 R20	Resistor, Var. (Beckman)	72P	680			R113			37.4 k ±	1 0.25 W
R21	Resistor (Mullard)	CR25	500 220	± 20 ± 5	0.5W 0.33W	R114 R115			432 k	1 0.25 W
R 22		-	10K	-"		R116				1 0.25 W 1 0.25 W
R24 R25] :			"	R117			174 k ±	1 0.25 W
R26			180 2.7K			R118 R119			143 k ±	1 0.25 W 1 0.25 W
R 27		"	2.2K	-		R120		"	110 k ±	1 0.25 W
R 28	Resistor, Var. (Beckman)	72P	2.0K	<u>+</u> 20	0.5W	R121 R130	Resistor	BTT	97.6 k ±	
R29 R30	Resistor (Mullard)	CR 25	220	± 5	0.33W	R131	"	BTT Code Z HOLCOH2	22 M ± 3.4 M ±	10 0.5 W 1 0.5 W
R31			10K 1.5K		-	R132		HOLCOH4	1.69M ±	
R32	* "		10K		-	R133 R134			l 1 13 Ml±	1 0 25 W
R33 R34			1.5K		:	R134 R135			845 k ± 681 k ±	1 0.25 W 1 0.25 W
R35		MR25	3.9K	<u>+</u> 2	0.4W	R136			562 k 🛨	1 0.25 W
R36	" "	· •	3.16K	"	٠.	R137 R138		"	487 k +	
R37 R38	Resistor, Variable	CR 25	3.3K	+ 5	0.33W	R139			374 k ±	0.25 W
R39	Resistor (Mullard)	A/E F3025 CR25	10K 3.3K	± 20 ± 5	0.75W 0.33W	R140 R141	••	BTT	340 k ± 22 M ±	1 0.25 W
R40		"	1.5K	± 5	"	R142		Code Z HOLCOH2	3.4M ±	1
R41 R42			330K			R143	•	Code Z HOLCOH4	1.69M ±	
R43		MR 25	10K 820	± 2	0.4W	R144	"	"	1.13M ±	1 0.25 W
R44	**		10K		•	, R145 R146		" "	845 k ± 681 k ±	0.25 W
R45		CR25	47	<u>+</u> 5	0.33W	R147	**	-	562 k 🛨	1 0,25 W
R46 ;	Resistor, (Var. (Beckman) Resistor, Code Z	72P HOLCOH4	500 340 k	± 20 ± 1	0.5W	R148 R149		- :	487 k ±	
R51			169 k	ži	0.25 W 0.25 W	R150	**		422 k ± 374 k ±	0.25 W 0.25 W
R52 R53		**	113 k	<u>+</u> 1 +1	0.25 W	R151 R152	••	, ,	340 k ±	1 0.25 W
R54		••	84.5 k 68.1 k	±1	0.25 W 0.25 W	R153	**	Code Z	22 M ±	1
R55 R56		:	56.2 k	<u>+</u> 1	0.25 W	R154	••	HOLCOH2 Code Z HOLCOH4	3.4 M ±	1
R57			48.7 k 42.2 k	±1 ±1	0.25 W 0.25 W	R155		HOTCOH	1.69M ±	
R58	* *		37.4 k	±1 +.	0.25 W	R156 R157			845 k 🛨	0.25 W
R59 R60		••	340 k 169 k	±1 ±1	0.25 W 0.25 W	R158			681 k ± 1	
R61			113 k	±1	0.25 W	R159 R160			487 k ± 1	0.25 W
R62 R63		.,	84.5 k 68.1 k	± 1 ± 1	0.25 W 0.25 W	R161	"		422 k + 1 374 k + 1	0.25 W
R64			56,2 k	+1	0.25 W	R162	**		340 k 21	0.25 W
R65 R66			48.7 k 42.2 k	±1 ±1	0.25 W 0.25 W	R163 R164	•	Code Z	22 M ± 1	
R67			37.4 k	±1	0.25 W	R165	•	HOLCOH4	665 k ± 1	0.25 W
R68 R69	H . H	:	340 k 169 k	± 1 ± 1	0.25 W 0.25 W	R166	••	"	442 k 1	0.25 W
R70			113 k	±1 ±1	0.25 W	R167 R168	"	"	267 k ± 1	0.25 W
R71 R72			84.5 k 68.1 k	±1 ±1	0.25 W 0.25 W	R169	**		221 k ± 1	0.25 W
R73			56.2 k	+ 1	0.25 W	R170 R171	*		187 k = 1 165 k = 1	0.25 W 0.25 W
R74 R75			48.7 k 42,2 k	±1 ±1	0.25 W 0.25 W	R172	··		147 k 🙏	0.25 W
R76			37.4 k	Ξī	0.25 W	R173 R174		втт	133 k 1 1 22 M 1 1	0.25 W 0 0.5 W
R77 R78			133 k 66.5 k	±1 ±1 ±1	0.25 W 0.25 W	R175	**	Code Z HOLCOH2	3.4 M = 1	
R79		-	44.2 k	1 ± 1	0.25 W	R176	•	Code Z HOLCOH4	1.69M ±1	
R80 R81	* *	::	33.2 k 26.7 k	±1 +1	0.25 W 0.25 W	R177			1.13 M ± 1	0.25 W
R82			20.7 k 22.1 k	±1 ±1 ±1	0.25 W	R178 R179	•		845 k + 1 681 k - 1	0.25 W 0.25 W
R83	·	. •	18.7 k 16.5 k	±1	0.25 W 0.25 W	R180			562 k + 1	0.25 W
R84 R85			10.5 k	±1 ±1 ±1	0, 25 W 0, 25 W	R181 R182		:	487 k - 1 422 k - 1	0.25 W 0.25 W
R86	** **			<u>+</u> 1	0.25 W	R183		-	374 k ± 1	0.25 W 0.25 W
R87 R88			169 k 113 k	±1 ±1	0.25 W 0.25 W	R184 R185		BTT	340 k 1 1	0.25 W
R89			84.5 k	±1 ±1	0.25 W	R186	•	Code Z HOLCOH2		
R90 R91			68.1 k 56.2 k	l±1	0.25 W 0.25 W	R187	•	Code Z HOLCOH4	3.4 M = 1 1.69M = 1	0.5 W
		:	48.7 k	±1 ±1	0.25 W	R188		"	1.13 M 1 ± 1	0.25 W 0.25 W
R92		f	42.2 k	1+1	0,25 W	R189		"	045 1 1+ 1	0 05 111
R93				Ŧ,			••		845 k 1	0.25 W
R93 R94 R95			37.4 k 340 k	 + :	0.25 W 0.25 W	R190 R191	• •		681 k + 1 562 k + 1	0.25 W
R93 R94			37.4 k	± 1 ± 1	0.25 W	R190		-	681 k ± 1 562 k ± 1 487 k ± 1 422 k ± 1 374 k ± 1	0.25 W

Circuit Ref.	1	Description	Ret.	Value	Tol.	Rating		Circu	uit
			No. Code Z	Ω	T			Ref.	4
R 195	Resisto	or	HOLCOH	4 340 k	± 1	0.25 W		X I	ļc
R196	5 "		BTT	22 M	± ic			X 2	- 1
	1		Code Z		- ^	10.0		Х3	-
R197	' "		HOLCOH	2 3.4 M	± 1	0.5 W		X4	ı
2100			Code 2		١.	i		X5	
R198 R199			HOLCOH	4 1.69 N		0.25 W		X6	
R 200				1.13 N	1 2 1	0.25 W		X7	
R201				845 k	Ξi	0.25 W		X8	
R 202				681 k	± 1 ± 1	0.25 W		Х9	
R 203	4		-	562 k	± 1	0. 25 W		X10	
R204			1 -	487 k	± 1	0.25 W		S2	S
R 205			-	422 k 374 k	Ξi	0.25 W 0.25 W		S3	- 1
R 206			-	340 k	1+.	0.25 W		. S4	
R 207	-		BTT	22 M] ± 1	0.5 W		\$5	- 1
	i		Code Z		1				
R 208	-		HOLCOH	2 8.66 M	±1	0.5 W			
R 209	"		-	4.32 M	±1	0.5 W			
R210	1 :		-	2:87 M		0.5 W			
R211	-		-	2.15 M	± 1	0.5 W		Filt	ler
D 21 2	_		Code Z	1	1				
R212 R213	"		HOLCOH			0.25 W		(H	ial
R213	-]	1.43 M	‡ <u>1</u>	0.25 W			Э,
R215	1 -		-	1.24 M	± 1 ± 1	0. 25 W			
R216	-		_	1.1 M 976 k	± 1	0.25 W		Circu	ait
	1		Code Z	7/0 K	-1	0. 25 W		Ref	
R217	-		HOLCOH4	866 k	±1	0.25 W		R22	
	1			1	1	J. 23 W		C15	
Cl	Canacino	r, Electrolytic	C 426	50.00	+ 50	25.11		C20	
C2	- packe	., Diectrosytic	AR/F50	50 μF	- 10	25 V		C25	
	1		TAG	1	1	I		C30	
C3	1 .	SOLID TANTALUM	0.1/35	01.00	± 20			C35	
C4	-	" " " "	1 2.27 33	0.1 μF 0.1 μF	± 20	35 V 35 V		C40	
C5	-	POLYSTYRENE	H.S.	υ.1 με	± 10	63 V		C45	
			TAG	1	I	1		C50	
C 6	-	SOLID TANTALUM	0.1/35	0.1 μF	± 20	35 V		C55	
C 7	-		1 5.2/33	0.1 μF		35 V 35 V		C60	- 1
C8	"	POLYSTYRENE	H.S.) · µr	± 10	63 V			
C9	"	*	H.S.	100 pF	± 10	63 V			
C10	"	CERAMIC	P100/YD	3.6 pF	I X 0.51	200 V			
CH	"	METAL. POLYESTE			± 10	400 V			
C21	"	**	H.S.	4700 pF	1 2 3 1	63 V			
C22 .	"	POLYCARBONATE	CTROIOC	0.047µF	± 2 ¯	160 V	. 1		
C23	1 :		CMD10C	0.47 μF	±2	63 V	· i	Filt	ter
C24	l .		CMD40C	4.7 μF	± 2	63 V	, 1	_	
C26 C27		POLYSTYRENE	H.S.	4700 pF	± 24	63 V	i •	(L	OV
C28	-	POLYCARBONATE "	CTR010C	0.047µF	+ 2	160 V	1	` _	_ •
C29			CMD10C CMD40C	0.47 μF	÷2	63 V			
C31	-	POLYSTYRENE	H.S.	4.7.µF		63 V	-	Circu	160
C32	-	POLYCARBONATE	CTROIOC	4700 pF 0.047 μF		63 V 160 V	1		•••
C33		*	CMD10C	0.47 μF		63 V	!	Ref.	
C34	"		CMD40C	4.7 uF	+2	63 V		C20	С
C36	"	POLYSTYRENE	H.S.	4700 pF		63 V		C25	
C37	"	POLYCARBONATE	CTR010C	0.047µF	± 2	160 V		C30	
C38	1 :	**	CMD10C	0.47 µF		63 V	į	C35	
C39	:		CMD40C	4.7 µF	+2	63 V		C40	
C41	:	POLYSTYRENE	H.S.	4700 pF	± 23	63 V	1	C45	
C42 C43	:	POLYCARBONATE	CTR010C	0.047µF	±2	160 V	į.	C50	
CAA			CMD10C		± 2	63 V			
C46	-	BOI VETURE	CMD40C	4.7μF	+ 2	63 V		C55	
C47	٠.	POLYSTYRENE POLYCAP BONATE	H.S.	4700 pF	<u> 구 2월</u>	63 V			
C48	-	POLYCARBONATE	CTR010C	0.047µF	¥2	160 V		*C5, (C8 æ
C49	-		CMD10C CMD40C	0.47μF	÷2	63 V		Value	
C51	۳ ا	POLYSTYRENE	H.S.	4.7 μF 4700 pF	± 2	63 V		•	
C52	•	POLYCARBONATE	CTR010C	0.047µP	7 52	63 V 160 V			
C53	-	"	CMD10C	0.47 µF		160 V 53 V			
C54	•		CMD40C	4.7 08	+ 2 1	53 V			
C56	•	POLYSTYRENE	H.S.	4700 pF		53 V			
C57	*	POLYCARBONATE	CTR010C	0.047µF	23 1	160V			
C58	-	*	CMD10C	0.47 uF	+ 2 6	53 V			
C59	. "	· · · · · · · · · · · · · · · · · · ·	CMD40C	4.7 μF		3 V			
DI :	Zener Di		YZ88/C5V6		1 -	.6V			
D2 .	••	" " "	"	1	<u>-</u>]]	.0 V			
D3	Diode (i.	.т.т.)	IN4148 NO	mp. na			lahanan		-4
D4		,,	IN4148 NO						
D5	**					D7 (EF3			17)
D6						tted if fo		seary	
D7	,	. [durin	g test	procedu	re.		
D8		1							

Circuit Ref.	Description		Ref. No.	Value Ω	Tol.	Rating
X 1	Operational	Amplifier	LM310N	i		
X 2	l "	•••				
X3	l "	••				
X4		••				
X5			-	İ		
X6		••				
X7	"	••				
X8	-	••	1321			
X9	-	••	LM310N	ł		
X10	-					
52	Switch		B/EF3020	ļ		
53	-		B/EF3021	ĺ		
54	•			1		
\$5	••		B/EF3022	1		

Filter Unit EF3-01 (High-Pass) only

Circuit Ref.	Description	Ref. No.	Value	Tol.	Rating
R220 C15 C20 C25 C30 C35 C40 C45 C50 C55	Resistor Capacitor, Polystyrene Capacitor,	MR 25 H.S. H.S. H.S. H.S. H.S. H.S. H.S. H.S	15 k Ω 10 pF 470 pF 470 pF 470 pF 470 pF 470 pF 470 pF 470 pF 470 pF	+ 2 + 10 + 22 + 22 + 22 + 22 + 22 + 22 + 22 + 2	0.4 W 160 V 63 V 63 V 63 V 63 V 63 V 63 V
		MD 306	10 μF	-3	63 V

Filter Unit EF3-02 (Low-Pass) only

				•			
Circuit Ref.	I	Description	Ref. No.	Value Ω	Tol.	Rating	_
C20 C25	Capacitor,	Polystyrene "	LCR	420pF	± 1	125V	
C30	•	••	••	•	**		
C35	••	••		••			
C40		••	••	390pF	•		
C45	**		•	" pr	**		
C50			•	**		••	
C55	**	••		••	**		

^{*}C5, C8 and Cll may be fitted if found necessary during test procedure. Values are given on capacitors.

DA15P 6P55670AH 32 7127 UG1094A/U ...

D8 PLi SK2

SK4 SK5 SK6 Plug, 15 Way Socket, Edge Connector

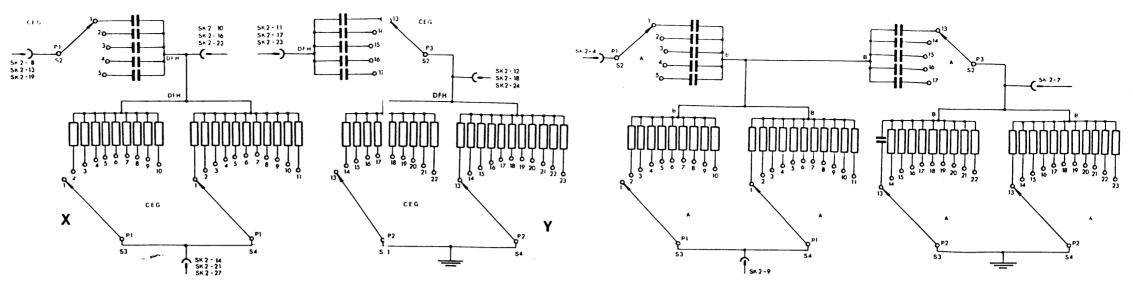
Coaxial Connector

Components Reference Table

	Switch	Pole	RC 1 From Wafer	RC 2 From Wafer	RC 3 From Wafer	RC 4 From Wafer
Switch	Positions	Pin	A to B	C to D	E to F	G to H
	1		C 24	C 34	C 44	C 54
	2	1	C 23	C 33 C 27	C 43 C 32	C 53 C 37
	3 4	Pl	C 22 C 21	C 26	C 31	C 36
	5		C 20	C 25	C 30	C 35
			1			
	7			1		
	8 9	P2	1			
	10	'-				
	11		ŀ	i		
S 2	13	 	C 29	C 39	C 49	C 59
	14		C 28	C 38	C 48	C 58
	15	P3	C 42	C 47	C 52	C 57 C 56
	16	1	C 41 C 40	C 46 C 45	C 51 C 50	C 55
	17	1	1	043		
	19					
	20	1			l	
	21 22	P4		1		
	23		i			
	1			ļ	 	
	1 2	1	R 50	R 59	R 68	R 77
	3	l	R 51	R 60	R 69	R 78
	4	1	R 52	R 61	R 70	R 79
	5	l	R 53	R 62 R 63	R 71 R 72	R 80 R 81
	6 7	Pl	R 54 R 55	R 64	R 73	R 82
	1 6		R 56	R 65	R 74	R 83
	9		R 57	R 66	R 75	R 84
	10		R 58	R 67	R 76	R 85
S 3	13	 	 	 	 	·
	14	1	R 86	R 95	R 104	R 113
	15	1	R 87	R 96 R 97	R 105 R 106	R 114 R 115
	16	1	R 89	R 98	R 107	R 116
	18	P2	R 90	R 99	R 108	R 117
	19	1	R 91	R 100	R 109	R 118
	20	1	R 92	R 101 R 102	R 110 R 111	R 119 R 120
1	21 22	1	R 93	R 103	R 112	R 121
		1	l	l	l	l
	1	T	R 130	R 141	R 152 R 153	R 163 R 164
	3	1	R 131 R 132	R 142 R 143	R 153	R 165
	1 4	1	R 133	R 144	R 155	R 166
	5	1	R 134	R 145	R 156	R 167
	6	1	R 135	R 146 R 147	R 157 R 158	R 168 R 169
	7	P1	R 136 R 137	R 148	R 158	R 170
	;	"	R 138	R 149	R 160	R 171
	10	1	R 139	R 150	R 161	R 172
	11	1	R 140	R 151	R 162	R 173
8 4	13	+	R 174	R 185	R 196	R 207
	14		R 175	R 186	R 197	R 208
	15		R 176	R 187	R 198	R 209
	16	1	R 177	R 188	R 199	R 210
	17 18	P2	R 178 R 179	R 189 R 190	R 200 R 201	R 211 R 212
	19	"	R 180	R 191	R 202	R 213
	20	1	R 181	R 192	R 203	R 214
	21	1	R 182	R 193	R 204	R 215
				. 9 104	R 205	R 216
	22 23	1	R 183 R 184	R 194 R 195	R 206	R 217

NOTES: (i) All wafers on common switch reference are mechanically coupled.
(ii) The component values are shown on Figure 7 (EF3-01 & EF3-02)
and Figure 10 (EF3-03 & EF3-04).

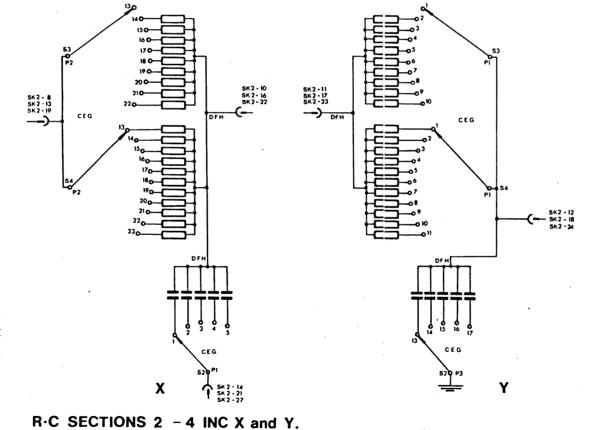
Filter Unit EF 3-01 & EF 3-03 (High-Pass)



R-C SECTIONS 2-4 INC X and Y.

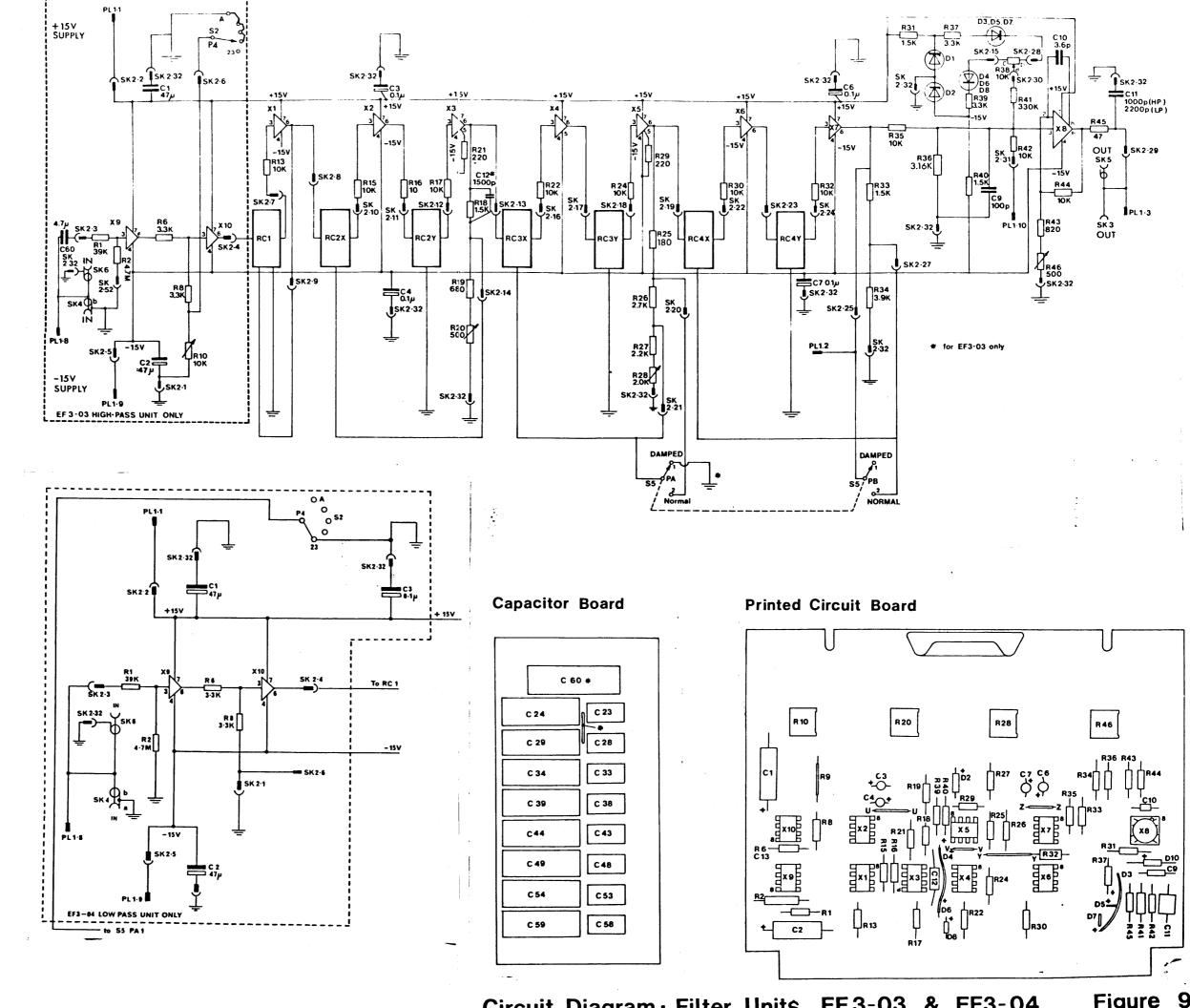
R·C SECTION 1.





R·C SECTION 1.

C23 C28. 6.47 MF



Circuit Diagram: Filter Units EF 3-03 & EF 3-04

Figure 9

	alue Ω	Tol.	Rating	Circuit Ref.	t	Description	Ref. No.	Value Ω	Tol.	Rating		Circu	lt	Description	Ref.	Value	Tol.	Rating	
	9K	+ 5	0.33W	R101	Resistor,	Holco	,5.H8	4.87K	<u>+</u> 1	0.25W		Ref.			No.	Ω			
	.7M	± 10	0.25W	R102		*		4.22K				R196 R197	Resistor, Resistor,		E.B. 'Z'H8	22M	± 10	0.25W	
	. 3K	± 2	0.4W	R103	•	•		3.74K			•	R198	"	***************************************	2 H8	340K 169K	± 1		
	# 01/			R104	-		-	34K	-	-		R199	•		-	113K	-	-	
10	OK "	<u>+</u> 5	0.33W	R 105 R 106	-			16.9K 11.3K				R 200	-	-		84.5K	••	-	
16	0Ω	-		R100	-	-	-	8.45K		-		R 201 R 202	-	-	*	68.1K		-	
	OK	-	••	R108	•	-	-	6.81K	-	-		R 202	•		•	56.2K 48.7K		-	
1.	.5K	-		R109	-			5.62K		-		R 204	-	-		42.2K		-	
	80		••	R110	-	-	-	4.87K	-	-		R 205	-	-		37.4K	-		
	00	± 20	0.5W	R111	-	-	-	4,22K		-		R 206		-	•	34K		•	
	20 0r	± 5	0.33W	R112 R113	-			3.74K 86.6K		-		R 207		A. Bradley	E.B.	22M	± 10	*	
10	OK		-	R113				43.2K		-		R 208	Resistor,	Holco	'Z'H4	866K	<u>±</u> 1	•	
11	80			R115	-			28.7K		-		R 209	-	-	'Z'H8	432K			
	.7K		**	R116	-		-	21.5K	**	-		R210 R211			-	287K		-	
	. 2K			, R117	-	•	•	17.4K	-	-		R211		-		215K 174K			
	.OK	<u>+</u> 20	0.5W	R118	-	-	•	14.3K				R213	•	-	-	1/4K 143K		-	
22	20	<u>+</u> 5	0.33W	R119	-	-		12.4K	-	-		R214	•		-	124K	-	•	
	0K		-	R120	-	-	-	11K	-	-	1 -	R215	•	-		110K	-	•	
	.5K	••	•	R121	n n/	A Dandless	 10 10	9.76K	± 10			R216	•	•	•	97.6K	•		
	OK .	-	-	R130		A. Bradley	E.B. 'Z'H8	22M 340K	± 10 ± 1	•		R217	•	**		86.6K		•	
	.5K	-		R131	Resistor,	110100	£ 110	169K	÷.	-	11:	Cl	Capacitor,	Blect. (Mullare	1) 016-16479	47µF	+50 -20	25V	
	,9K		 0.4W	R132 R133				113K		•	ļ	C2	-		•		-20	•	
	DK . 16K	± 2	0.4W	R134				84.5K		-		C3	Capacitor	Tant (L. T. T.)	TAGO-1/35	0. lµF	<u>+</u> 20	35V	
		45	0. 33W	R135	-	-	•	68.1K		•		C4		" '	., .,	"	- :		
	OK	± 5 + 20	0.75W	R136	-		-	56.2K	•	-		C6	•			•	**	-	
	.3K	± 20 ± 3	0.33W	R137	-		-	48.7K	•	-		C7	•		**	•	-		
	. 5K	± 5		R138	-	*		42.2K		-		C9		P'styr. (I.T.T		100pF	± 10	63V	
	30K			R139	•			37.4K	-	-		C10		Ceramic (L.T.		3.6pF	± ₫,	200V	
		<u>+</u> 2	0.4W	R140	-		#·	34K				C21		P'styr. (Suflex		4700pF		30V	
	20			R141		, A. Bradley	B. B.	22M 340K	± 10			C22 C23	Capacitor	r carb. (Adv. I	CMD10C	.047µF		160V	
	OK			R142	Resistor,	, 1101C0	'Z'H8	169K	± 1		- 11	C23	# ·		CMD10C CMD40C	.47μF	± 2	63V	
47		± 5	0.33W	R143				113K				C24		P'styr. (Suflex)		4.7μF 4700pF	± 2 + 21	30V	
		± 20	0.5W	R144 R145	-	••		84.5K		•		C27			'cap)CTR010C	•700pr	+2	160V	
	4K Kor	± 1	0.25W	R145	-		•	68.1K	**			C28		" "	CMD10C	.47μF	- <u>:</u>	63V	
	6.9K 1 9K			R147	-	•		56, 2K	•	-		C29			CMD40C	4.7μF	•		
	1.3K .45K			R148	-	•		48.7K	•		 	C31	•	P'styr. (Suflex)		4700pF	+ 21	30V	
_	.81K			R149	-	•	•	42, 2K	**	•		C32		P'carb. (Adv. F'		.047µF	± 2	160V	
	.62K	**	••	R150	-	•	•	37.4K	**	•		C33	•		CMD10C	.47µF		63V	
	87K			R151	-	**		34K		-		C34	-	* *	CMD40C	4.7μF		*	
	22K	•		R152		, A. Bradley	E.B.	22M	± 10	-		C36		P'styr. (Suflex)	H.S.	4700pF	± 2½	36V	
	74K	**		R153	Resistor	, Holco	'Z'H8	340K 169K	± 1			C37 C38		r card. (Adv. F	'cap) CTR010C	.047µF	± 2	160V	
	4K	**		R154	-			109K 11 3 K	••			C39			CMD10C CMD40C	.47μF 4.7μF		63V	
	5.9K			R155 R156	-			84.5K	••	•		C42		P'carb. (Adv. P	cmD40C	•. /μr •047μF		160V	
	1.3K			R157	-			68.1K	••	•	' I	C43		" "	CMD10C	.47μF	-	63V	
	.45K .81K	**		R158	-			56, 2K		•		C44			CMD40C	4.7µP	•		
	.62K	••		R159	-	••	•	48.7K	•	•		C47	•		CTR010C	.047µP		160V	
	.87K			R160	-	-	•	42, 2K	*	-		C48	•		CMD10C	.47µP	•	63V	
	. 22K	**	•	R161	•			37.4K	••	-		C49			CMD40C	4.7µF		•	
	74K	**	•	R162	_			34K				C52		P'carb. (Adv. P		.047µF	•	160V	
34	4K	•		R163		, A. Bradley	E.B.	22M	± 10			C53			CMD10C	.47µP		63V	
	5.9K	•	*	R164		Holco	'Z'H8	133K	±1			C54		- "	CMD40C	4.7μF			
	1.3K		••	R165				66.5K 44.2K				C57 C58			CTR010C	.047µF		160V	
	.45K	**		R166	•	*		33, 2K				C58			CMD10C CMD40C	.47µF		63V	
	.81K			R167 R168	-			26.7K				DI	Zener Dia	ie (Muliard)	BYZ88/C5V	4.7μF 6	+5	5.6V	
	.62K			R169	-			22. 1K			1	D2		"	2 1 200/C3 V	-	± 5		
	, 87K , 22K	••		R170		••	•	18.7K		•		D3	Diode (I.7	r.T.)	IN4148 NO	TE: D3 r	o D7 (E.F	3-03 high-pass uni	t) and
	. 74K			R171		•	••	16.5K	•	•		D4	" "	-•	" 10			7 (EF3-04 low-pass	
	3.3K	••		R172			. ••	14.7K	•	•		D5						d if found necessar	
	.65K			R173				13.3K		•	1	D6			• ,			rocedure,	•
	.42K		•	R174		r, A. Bradley	E.B.	22M	± 10			D7			•	i			
	32K		*	R175		r, Holco	'Z'H8	340K	± 1	-	4	D8			•				
	.67K	••		R176		-	-	169K	-		! }	PL1	Plug, 15 W		DA15P				
2.	.21K	*	**	R177		-	-	113K		7	. 1	SK2	Socket, Ed	ge Connector	6P55670AH	:			
	.87K			R178				84.5K 68.1K		•			6		32 7127				-
	.65K	**		R179			••	56.2K			1	SK3	Coexial Co	emector	UG1094A/U				
	.47K			R180 R181				48.7K			1	SK4 SK5							
	K Or			R181				42.2K				SK6							
	6.9K 1.3K			R183				37.4K			l	XI	Operations	l Amplifier	LM310N				
	1. 3K . 45K	-		R184		•		34K	•	•	-	X2	a her actoria	*					
	. 81K	-		R185		r, A. Bradley	E.B.	22M	± 10			X3	•	-	-				
	.62K	-		R186		r, Holco	'Z'H8	340K	<u> </u>	•		X4	-	•	-				
	.87K			R187	-	•	•	169K		-		X5	-	-	-				
	. 22K		•	R188	-	•	-	113K		.		X6		•	•				
	.74K	•	•	R189		•	.	84.5K		-	-	X 7	•	•	-				
34	4K	•		R190		•		68.1K				X8	-	•	1321				
	5.9K		•	R191		-	-	56.2K				X9		-	LM310N				
	1.3K		•	R192			-	48.7% 42,2%		•	ì	X10	•	•					
	45K	-	-	R193			-	37.4K		•	1	S2	Switch		B/EF3020				
			-									S3			B/EF3021				
	.81K .62K	_		R195	•	-	-	34K	-			S4	-						

Circuit Ref.		Descrip	tion	Ref. No.	Value Ω	Tol.	Rating
R10	Resistor,	Var. (Be	ckman)	72P	10K	+ 20	0.5W
R220	Resistor,			CR 25	330	- 5	0.25W
Cll	Capacitor,	P'styr.	(L.T.T.)	H.S.	1000pF	+ 10	63V
C12	Capacitor,	, P'styr.	(Suflex)	H.S.	1500pF		160V
C20	Capacitor,	P'styr.	(L.C.R.)		470pF	+ 2	3 0V
C25	•	*	-				•
C30	•		•		-	••	•
C35	•		-		-	•	••
C40		-				**	•
C41	**		-		4700pF	••	
C45	-	•	-		470pF		•
C46			-		4700pF	*	
C50			-		470pF	••	
C51			•		4700pF	•	•
C55		-	-		470pF	••	
C56	••	-			4700pF		•
C60	•	P'carb.	(Adv. F'ca	p)CMR40A	4.7µF	<u>+</u> 10	160V

Filter Unit EF3-04 (Low-Pass) only

Circui Ref.	t 1	Descrip	nion "	Ref. No.	Value Q	Tol.	Rating
Cll	Capacitor,	P'styr.	(L T. T.)	H.S.	2200pF	+ 10	63V
C20	Capacitor,	P'styr.	(L.C.R.)		420pF	- 2	30V
C25							*
C30					•		
C35							
C40		-	*		390pF		
C41	н	**	**		4500pF		
C45	••	-	*		390pF	•	
C46					4500pF	••	
C50	••	*			390pF	**	
C51		**			4500pF	••	
C55			••		390pF		**
C56	••	••	**		4500pF	••	

R92 R93 R94 R95 R96 R97 R98 R99 R100

Circuit Ref.

R 19

R22 R 24 R 25 R 26 R 27 R 28

R30 R31 R32

R33 R34 R35 R36

R42 R43

R1 Resistor (Mullard) R2 Resistor (Dubilier) R6 Resistor (Mullard)

R20 Resistor, Var. (Beckman)

Resistor, Var. (Beckman) Resistor (Mullard)

R45 " " R46 Resistor, (Var. (Beckman) R50 Resistor, Holco

